









$C_I$  is the I channel conversion loss factor,  $C_Q$  is the Q channel conversion loss factor,  $\theta_R$  is the I-axis phase rotation in radians,  $B_I$  is the I channel DC offset in volts,  $B_Q$  is the Q channel DC offset in volts, and  $\theta_E$  is the quadrature phase error in radians.

When the LO and RF frequencies are not equal,  $\theta_R$  can be set to 0 to simplify (5) and (6):

$$\hat{I}(t) = C_I I(t) + B_I \quad (7)$$

$$\hat{Q}(t) = C_Q (\cos \theta_E Q(t) - \sin \theta_E I(t)) + B_Q \quad (8)$$

$\theta_R$  is only important in applications when the phase difference between the RF and LO signals must be known (i.e. phase detector).

**Example:** Apply a 10 GHz CW LO signal at +5 dBm and a 10.001 GHz CW RF signal at -2 dBm. To estimate the AD90120B's  $\hat{I}(t)$  and  $\hat{Q}(t)$  signals, start by determining all the parameters in (7) and (8).

$C_I$  and  $C_Q$  are determined by the conversion loss and amplitude imbalance of the AD90120B. From the datasheet's typical performance plots at 10 GHz, use 8 dB conversion loss and -0.12 dB amplitude imbalance to find  $C_I$  and  $C_Q$  :

$$\frac{C_I + C_Q}{2} = 10^{(-8/20)} = 0.3981 \quad (9)$$

$$20 \log\left(\frac{C_Q}{C_I}\right) = -0.12 \quad (10)$$

$$C_I = 0.4008 \quad C_Q = 0.3954 \quad (11), (12)$$

Quadrature phase error and DC offsets are also obtained from the typical performance plots at 10 GHz:

$$\theta_E = 0.1 \text{ Deg.} = -0.0017 \text{ Radians} \quad (13)$$

$$B_I = -0.0045 \text{ V} \quad B_Q = -0.0062 \text{ V} \quad (14), (15)$$

The next step in estimating  $\hat{I}(t)$  and  $\hat{Q}(t)$  is to calculate the ideal  $I(t)$  and  $Q(t)$  from the RF input signal. Given that the RF signal frequency is 1 kHz greater than the LO frequency,  $I(t)$  and  $Q(t)$  define an upper sideband tone of 1 kHz having a constant amplitude of:

$$\frac{A^2}{0.1} = 10^{(-2.0/10)} \quad (16)$$

$$A = 0.2512 \text{ V} \quad (17)$$

From (3) and (17) we know:

$$I(t) = 0.1776 \cos(2\pi 1000t) \quad (18)$$

and

$$Q(t) = 0.1776 \sin(2\pi 1000t) \quad (19)$$

The final step in estimating  $\hat{I}(t)$  and  $\hat{Q}(t)$ , the demodulator's real I and Q outputs signals, is to insert (11), (12), (13), (14), (15), (18), and (19) into (7) and (8) giving the final result:

$$\hat{I}(t) = 0.071 \cos(2\pi 1000t) - 0.0045$$

$$\hat{Q}(t) = 0.070 \sin(2\pi 1000t - 0.0017) - 0.0062$$